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Miniaturized and Flexible FSS based EM-Shields for Conformal Applications

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Abstract— This paper reports on very efficient and highly miniaturized wideband Polygon shaped Frequency Selective Surface (PFSS) shields for planar and conformal applications in X band. These shields have been analyzed for both lossy as well as for low loss substrates. The conformal configurations of PFSS are particularly investigated for inward, outward and double curved profiles useful for a variety of applications. The conformal designs have been tested for two different radii of curvatures. The models have been simulated using a hybrid simulation approach for electrically large geometries. Equivalent circuit model as well as analytical model is determined. The proposed PFSS designs offer stable angular response up to 60° for planar and all conformal geometries on both lossy and low loss prototypes. Shielding effectiveness of at least 55 dB and 48 dB has been measured for flat/non conformal and conformal configurations.

Index Terms— Bandstop filters, Electromagnetic compatibility, Frequency Selective Surface (FSS), Interference, Shielding, Polarizations.

I. INTRODUCTION

FREQUENCY selective surfaces (FSS) have been exhaustively investigated since the 1960s. An FSS is combination of a periodic structure designed to selectively absorb/reflect/transmit electromagnetic waves [1]. These surfaces can be referred to as spatial filters i.e. they can respond to a particular frequency in the whole electromagnetic spectrum [2]. Recent research trends lay emphasis on FSSs owing to their potential applications in; satellite communication services [6], antenna radomes [3], Radar Cross Section (RCS) reduction [4], efficient communication systems [5], performance enhancement of antennas, dichroic sub-reflectors, analog absorber and many others.

The performance of broadband communication systems that operate on low power spectral density and in particular portable communication devices are badly affected by electromagnetic interference (EMI). An efficiently designed FSS is a potential solution which not only mitigates interference but also maintains transparency to the rest of useful frequency spectrum. Planar FSS geometry can only be applied to flat surfaces thus various curved or pointed objects like cylinders or cones cannot be shielded. These kinds of applications demand for a conformal FSS shield which retains its performance when conformed around the desired object.

In recent literature, various planar and conformal FSS geometries are reported. In [8], A broadband flexible textile-integrated band stop FSS is presented. This large FSS achieves stable shielding response with a limited conformability. A flexible low profile Convoluted Ring Loop (CRL) FSS printed on Polyethylene Terephthalate (PET) is realized in [9] to achieve shielding at 10 GHz. This FSS achieves stable incident angular response for both Transverse Electric (TE) and Transverse Magnetic (TM) wave modes but up to 45° only. A dual band U-shaped FSS, reported in [10], effectively achieves EM shielding for X and Ku bands. In [11], a single-layer absorber containing two-dimensional array of cross dipoles are presented for RCS reduction. Another work shows band-stop FSS having a cross spiral shaped patch along with iterated H-shaped fractals arranged in the C-4 rotation to achieve polarization independent band-stop response [12]. In another study, a method of achieving wide range of beam widths by incorporating an outward curved conformal FSS in an antenna is presented [13]. Meandering technique makes this FSS significantly miniaturized and stable. In [14], flexible double-sided linear-to-circular polarizers are realized on polydimethylsiloxane. This configuration is based on two different unit cells, a loaded circular patch unit cell or an unloaded circular patch unit cell, both backed by a generic rectangular element. This double-sided polarizer achieves good phase difference, axial ratio (AR) and overall stable transmission response by employing inward and outward curved configurations. A couple of narrowband dielectric FSSs is reported in [15] to achieve reflection and transmission characteristics. In [16], a complementary symmetry is proposed to provide band-pass response. This complementary X-shaped FSS offers stable incident angular response along with miniaturized dimensions on a thin substrate. Electromagnetic shielding in X-band is achieved in [17] by realizing a convoluted square loop FSS on planar and curved flexible laminate. This analysis only investigates outward curved FSS surfaces. In [18], a miniaturized metamaterial based FSS is reported. This FSS is realized on a flexible ferrite laminate for generic microwave applications. A simple

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outward curved profile is investigated as an application. An active FSS is reported in [19] for beam switching applications. These slots based active FSSs are realized on a flexible substrate that circularly warps the dual band antenna for beam switching. In [20], a meandered loop based printed FSS is reported. The meandered profile of this FSS is optimized to design Radio Frequency (RF) planar shield for windows. The reported FSS offers optimum shielding while maintaining a stable incident angular response. Another single layer and dual band FSS design study is reported in [21]. The cross and ring shaped FSS are etched in a back-to-back planar configuration to achieve wide band shielding over limited angular stability. Another study reported in [22] reveals super formula-based creation of a dual band FSS where only planar configurations are desired. The super formula based square loop and inner ring achieves stable incident angular response while maintaining a miniaturized profile. In [23], a serpentine strip FSS based flexible shield is reported for flexible electronics. This flexible FSS shield incorporates all the design constraints while offering optimum shielding effectiveness when conformed only in inward configuration. Three ring slot pairs are reported in [24] for Ka-band radome applications. This reported FSS based radome offers flat passband response over a limited angular stability. In [25], a hybrid dual band FSS is reported. This configuration comprises of back-to-back arrangement of hexagonal patch in combination with hexagonal ring and Jerusalem cross. The reported design offers a narrowband reflective and transmissive response over Ku-band satellite and Ku-band active radars respectively. An ultra-wide stop-band single-layer FSS is reported in [26] to enhance the gain of an Ultra-Wideband (UWB) monopole antenna. The unit cell of the reported FSS consists of four asymmetric rectangular patches with circular slots embedded in it. This concept of four slotted patches is conceived to achieve ultra-wide stop-band characteristic with limited angular stability. In [27], a substrate integrated waveguide (SIW) technology based FSS is reported for airborne radome applications. This FSS consists of a tapered cross-slot on either side of the substrate surrounded by metallic vias. The structure selectively allows the impinging electromagnetic (EM) wave through it in the specified frequency band. The element shows very stable frequency response for oblique incidence and the narrow band roll-off performance characteristics at the edges of operating region in the stop band.

In the available literature, narrow band performance, lack of miniaturization, design complexity and incidence angular stability are the major limitations to overcome alongside. For conformal applications generally outward curved geometry is investigated.

The Polygon shaped Frequency Selective Surface (PFSS) proposed in this study addresses almost all the design constraints related to miniaturization, conformality and incident angle responses. The proposed PFSS effectively mitigates interference in X-band while maintaining stable incident angular performance up to 60° in both polarization states. Planar and conformal shields including outward, inward, double curved configurations are designed and analyzed through the simplistic, quick and computationally less intensive full wave Finite Element Boundary Integral (FE-BI) system model [7]. Equivalent circuit model, analytical model and full wave analysis has been carried out to validate the shielding performance of the proposed PFSS.

Rest of the paper is organized as follows; geometrical configurations and evolution of the proposed flexible PFSS are presented in Section II. Design modeling and equivalent circuit model has been reported in Section III. Section IV presents parametric analysis and optimization of PFSS unit cell. Section V briefly explains the simulation setup for planar and conformal configuration. Measurement setup and measured results of shields are presented and compared in Section VI. Finally, Section VII concludes the paper.

II. GEOMETRICAL CONFIGURATION

The unit cell design configuration of PFSS is presented in Fig. 1(a). This PFSS is simulated, fabricated and measured on two different substrates. For planar and non-conformal shield FR4 material ($\epsilon_r = 4.4, \tan\delta 0.02$) is used to analyze the PFSS behavior in lossy substrate. For conformal shields Rogers 5880 ($\epsilon_r = 2.2, \tan\delta 0.0009$) flexible laminate has been used. The periodic arrangement of the PFSS is illustrated in Fig. 1(b). The inter-element spacing of unit cells is optimized and fixed to $D_e = 0.3\ mm$ for both variants considering the fabrication limitations. Perspective view of the proposed FSS is presented in Fig. 1(c).

The proposed FSS geometry is mainly evolved through a 4-step design procedure as presented in Fig. 2. The unit cell comprises of twelve-sided polygon geometry. The total perimeter of the polygon is about $\lambda_0$ at the resonant frequency as shown in Fig. 2a. An inflated X-shaped geometry having center width $d_{1}$ is etched in the center of polygon initially. This inflated X-shaped geometry mainly helps to achieve band-stop behavior at the desired frequency band as illustrated in Fig. 2b. It may be observed from the Fig. 2c that, the center width of inflated X-shaped geometry is further increased and optimized to achieve band-stop behavior at X-band centered at 10 GHz. The inner corners are rounded with radii $d_3$ to improve its angular repose and enhance bandwidth as shown in Fig 2d. The optimized geometry offers miniaturization as well as polarization independence owing to its symmetrical profile. The optimized dimensions including their optimization ranges are listed in Table I and Table II for both variations of design realized on Rogers 5880 and FR-4 laminate respectively.
Fig. 1. Layout of PFSS. (a) design configuration (b) array arrangement (c) perspective view

Fig. 2. Design procedure of the proposed PFSS

<table>
<thead>
<tr>
<th>Table I</th>
<th>Dimensions of PFSS (Rogers 5880)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Variables</td>
<td>Optimization range (mm)</td>
</tr>
<tr>
<td>Length × width</td>
<td>7.5-8.5</td>
</tr>
<tr>
<td>(d_1)</td>
<td>3-4.5</td>
</tr>
<tr>
<td>(d_2)</td>
<td>1.5-2.6</td>
</tr>
<tr>
<td>(d_3)</td>
<td>1.5-3</td>
</tr>
<tr>
<td>(S_1)</td>
<td>2-3</td>
</tr>
<tr>
<td>(S_2)</td>
<td>1.5-3</td>
</tr>
<tr>
<td>(r)</td>
<td>2.8-3.4</td>
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<td>(th)</td>
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</tbody>
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III. Simulations and Modeling

Complete designing and optimization of the proposed PFSS is carried out in a commercially available 3D full wave electromagnetic simulator (Ansys HFSS®). The simulator solves Maxwell equations at the boundaries of tetrahedral mesh of the geometry. Floquet theory-based simulation setup was used for planar configuration with periodic boundary conditions whereas FE-BI simulation setup was used for conformal variants, more details on simulation setup is presented in Section V. Shielding Effectiveness (SE) of the proposed PFSS is expressed in (1), [17].

\[
SE_{dB} = -20 \log \frac{E_{out}}{E_{in}}
\]

The surface current density (J-surf) of this FSS array is examined at the resonant frequency of 10 GHz as shown in Fig. 3. The distribution reveals that current is more concentrated around the design corners. The significance of the inflated X geometry introduced in middle of the polygon is clear from this current distribution. The path of current can be increased or decreased by varying the shape without changing the overall size of polygon. The current intensity on metal introduces inductive effect, while the central X shape opening contributes to the capacitance.
An equivalent circuit model (ECM) of the proposed PFSS is developed to provide a physical interpretation of its performance determined by the full-wave simulations. The derivation of ECM mainly depends on the geometry of an FSS structure and the polarization of incident wave [22]. The surface current distribution on the proposed unit cell at 10 GHz under normal wave incidence (TEz-0°), also provide clues on the inductance and capacitance components in the equivalent circuit.

As a result, capacitances \( C_1 \) and \( C_2 \) are formed at the metallic gaps between PFSS unit cells and inflated X-shaped region. The inductances \( L_1 \) and \( L_2 \) are formed at the outer metallic edges of the PFSS as shown in Fig. 4a. The resultant ECM formation of the this FSS is presented in Fig. 4b. It may be observed from the Fig. 4b that, \( L_1 \) and \( C_1 \) are connected in series while parallel combination is visible between \( L_2 \) and \( C_2 \). The values of inductances and capacitances depend on the dimensions of FSS structure.

The free space on both sides of the unit cell is modeled as a transmission line with characteristic impedance of \( Z_0 = 377 \Omega \). The characteristic impedance of substrate is \( Z_{sub} = \frac{Z_0}{\sqrt{\varepsilon_r}} \). The substrate thickness \( h \) and material characteristics affect the phase difference of incident wave [22].

The value of components of the equivalent circuit is determined by the use of final ECM transmission line model. The design analysis also reveals that due to the looping nature of geometry the increment in \( L \) and \( C \) lowers the resonant frequency. The proposed circuit model achieves band stop behavior at 10 GHz and its impedance is derived in (2).

The derived expression is categorized into two zeros and two poles. The relations for the lumped parameters are obtained by equating both the numerator and denominator of (2) equals to zero.

\[
Z_{PFSS} = \frac{[\cos^2(c_{12}) - (1 - \cos^2(c_{12})/(1 - \cos^2(c_{12})))]}{\cos(c_{12})/(1 - \cos^2(c_{12}))}
\]

\[
C_2 = \frac{1}{2\omega_p^2}
\]

\[
C_1 = \frac{\omega_p^2 - \omega_z^2}{-2\omega_p^2\omega_z^2}
\]

In the given equations (3), (4) and (5), \( \omega_p \), \( \omega_z \) and \( \omega_z^2 \) are the pole and two zeros respectively. Whereas, the first pole \( \omega_{p1} \) is zero. The numerical value of \( L_2 \) has been inserted iteratively to evaluate rest of lumped parameters. The lumped parameters listed in Table. III, are calculated by using the above-mentioned relations.

\[
L_1 = \frac{(\omega_p^2 - \omega_z^2) - C_1 L_2 \omega_p^2 \omega_z^2}{C_1 \omega_z^2 (\omega_p^2 - \omega_z^2)}
\]

The SE can be directly plotted using the equivalent circuit model shown in Fig 4 through an EDA tool. Alternatively, the transmission coefficient of the proposed PFSS can be approximated by transmission line theory (6), after evaluating the lumped parameters [28].

\[
S_{21} = \frac{Z_2 Z_0 + B Z_0 + D Z_0}{AZ_0 + B + CZ_0 + D Z_0}
\]

where coefficients: \( A, B, C \) and \( D \) refers to the \( ABCD \) matrix of the system having \( N \) number of FSS dielectric layers.

\[
[A \ B] = [M_{PFSS}] [M_{n+1}] \ldots [M_N]
\]

In 7, \( M_n \) is the scattering matrix of nth dielectric layer. Here in this particular design we have one dielectric layer on which the FSS design is resting thus the relation (7) is solved for single layer and it reduces to (8)

\[
[M_s] = \begin{bmatrix} 1 & 0 \\ 1/Z_{PFSS} & 1 \end{bmatrix}
\]

Same procedure can be adapted for solving the TE and TM wave modes where \( Z_{PFSS} \) in (10) is replaced with \( Z^{TE} \) and \( Z^{TM} \) as given in (11) and (12) respectively.

\[
Z^{TE} = \frac{(\omega_p \mu_0 p_0)}{k_{z1}}
\]

\[
Z^{TM} = \frac{k_{x1}}{(\omega_0 \varepsilon_0)}
\]

Where \( k_{z1} \) and \( k_{x1} \) are normal and tangential components of the wave number respectively and are given in (13) and (14). \( \mu_0 \), \( \mu_0 \) and \( \varepsilon_0 \), \( \varepsilon_0 \) are free space and relative magnetic permeabilities, and the free space and relative dielectric permittivities. Free space propagation constant is represented by \( k_0 \) and \( d \) represents the thickness of dielectric. The
incident angle of the wave striking the FSS surface is given by $\theta$.

$$k_{z1} = \sqrt{\varepsilon_r \varepsilon_0 - k_l^2}$$  \hspace{1cm} (13)

$$k_l = k_0 \sin \theta$$  \hspace{1cm} (14)

Comparison of SE simulated through unit cell Floquet’s model along with SE simulated from circuit model and analytical model is presented in Fig 5. Excellent coherence can be observed in simulated and computed response.

IV. PARAMETRIC ANALYSIS AND DESIGN OPTIMIZATION

A. Geometric analysis

The resonant frequency and SE are tuned by parametric optimization. Initially, the perimeter of the unit element is approximated at $\lambda_0$. Afterwards the optimum results have been achieved by tuning other parameters while maintaining the desired SE. The fine-tuning parameter of PFSS is the width of inflated X slot $d_{L1}$. It may be observed from the Fig. 6(a) that resonant frequency shifts to lower frequencies as central width $d_{r1}$ increases by reducing the metallic surface area. The radius $r$ and the gap $d_{r2}$ of each corner are varied alongside to optimize the SE. The reduction in $r$ and $d_{r2}$ shifts resonant frequency towards higher frequencies as illustrated in Fig. 6(b). The thickness effect on SE of the employed laminate is also investigated in Fig. 6(c). It may be noticed that, the reduction in thickness, shifts the resonant frequency to higher frequencies. Overall, SE of not less than 55 dB is observed for PFSS while maintaining miniaturized dimensions and wideband response.

B. Substrate Analysis

The relative permittivity of the employed laminate is varied to investigate its effect on the performance of PFSS. It may be observed from the Fig. 7 that; resonant frequency tends to shift from higher to lower frequencies as $\varepsilon_r$ increases. This is true since the electrical length of resonant element is inversely related to the square root of dielectric constant.

The dielectric loss tangent $\tan \delta$ of employed laminate is also varied to study its effect on overall shielding response. As shown in Fig. 8, SE exponentially decreases as the loss tangent increase from 0.009 to 0.5. It happens because the resonant elements fail to efficiently resonate at the desired frequency due to higher loss.

V. SIMULATION SETUP FOR PLANAR AND CONFORMAL DESIGNS

For planar designs the periodic boundary conditions can be employed therefore the incident angular performance of the proposed unit cell of PFSS is investigated by employing Floquet's method. A unit cell is assigned periodic boundary conditions and it is excited though a Floquet port to simulate the infinite periodicity. The SE is simulated up to an angle of
for both TE and TM modes. It is also worth mentioning that the element separation is much smaller than operating wavelength so only fundamental Floquet mode shall exist at the operating frequency. It can also be ensured from the relation (15). Where $\lambda_{g}^{fr}$ shows the wavelength of the first higher order harmonic of Floquet mode and $\theta$ is the incident angle.

$$\lambda_{g}^{fr} = D_{c}(\sqrt{\varepsilon_{eff} + \sin \theta}) \quad (15)$$

The conceptual view of conformal configurations is presented in Fig. 9. The simulation setup of conformal shield is not straightforward since the field distributions are not symmetric on the left/right and top/bottom boundaries of a unit cell therefore Floquet method cannot be applied. A conventional free space simulation setup can be used but owing to electrically large geometry the computations complexity will increase multifold. Therefore, the conformal configurations have been analyzed by employing hybrid full wave FE-BI system model. Full wave simulation setup for analyzing all conformal PFSS shields is presented in Fig. 10. Each shield contains $25 \times 25$ unit elements and is placed between transmitting and receiving horn antennas. These horns are designed and optimized to operate in X-band (8 – 12 GHz). FE-BI boundary conditions are assigned to PFSS array/shield and both horn antennas. Owing to the closer boundaries, the overall free space region is reduced which results in faster simulation [7]. These full wave simulations took about 14 hours with 18 GB of RAM on Xeon E5-1650 hexa core CPU clocked at 3.5 GHz.

Fig. 10. FE-BI simulation setup for PFSS double curved shield

VI. MEASUREMENT SETUP AND RESULTS

Full wave simulated results of planar and conformal PFSS shields are validated by employing free space measurement setup illustrated in Fig. 11. A $7.5\lambda_{0} \times 10\lambda_{0}$ planar and conformal PFSS shields are fabricated on FR-4 and flexible Rogers 5880 with thickness of 1.6 mm and 0.127 mm respectively as illustrated in Fig. 12. An aluminum wall with an aperture in the middle has been used for measurements. These shields are fixed in the aperture one-by-one so that all radiations only pass from the shield. A pair of Lucas Nuelle WR90 15 dBi horn antennas operating in X-band is placed in the far field on both sides which are connected to Agilent PNA N5242A network analyzer. The employed measurement setup is initially calibrated to normalize the reference transmission level through the empty aperture.

Fig. 11. Shielding measurements of PFSS shield on flexible laminate (a) non-conformal / flat PFSS shield under measurement. (b) outward curved PFSS shield (c) inward curved PFSS shield (d) double curved PFSS shield under measurement.

Once the calibration was completed planar and conformal PFSS shields were installed in the aperture between these antennas to measure their shielding effectiveness. Multiple
measurements were performed to check variability/uncertainty (δ) in the measured results [29]. Overall, δ = ± 0.06, δ = ± 0.05, δ = ± 0.08 and δ = ± 0.1 is observed for flat, outward, inward and double curved shield measurements respectively. This slight variation in measured results may be observed due to sensitivity of FSS for small changes, curved profile, fabrication imperfections and possible errors in measurement setup.

B. PFSS shield realized on Rogers 5880

The SE of the low loss laminate based PFSS shield is compared for simulated and measured responses as shown in Figs. 14(a) and 14(b). The FSS shield achieves SE of not less than 55 dB. However, 2% variation is observed in measurements as the angle of incidence increases up to 60° as illustrated in Figs. 14(c) and 14(d).

C. Conformal Configurations

The proposed PFSS shield is conformed in three configurations for conformal applications. The testing setup for conformal configurations for normal as well as oblique angles is presented in Figs. 11(b), 11(c) and 11(d). Calibration is again performed prior to testing of all conformal shields.

1) Outward curved PFSS shield

The shielding response of outward curved shield is evaluated and compared with full wave simulations by varying the radius of curvature $R_o$ (Fig 9a). The shielding behavior of the proposed outward curved PFSS shield is presented in Fig. 15. It may be observed that, SE is slightly reduced to 50 dB and shifts only 1% and 2% from the resonant frequency of 10 GHz, as the radius $R_o$ reduces from 12cm to 6cm respectively. The incident angular performance of this conformal configuration is illustrated in Figs. 15(c) and 15(d). The oblique performance is evaluated by varying $\theta$ up to 60° at both variations of $R_o$. The SE shifts only 2% as the angle of incidence increases for both TE and TM-wave mode at both variations of $R_o$. Overall, this configuration exhibits stable incident angular behavior in all variations.
2) **Inward curved PFSS shield**

A good agreement between simulated and measured shielding performance is observed for the proposed inward curved PFSS shield as shown in Figs. 16(a) and 16(b). This curved configuration also offers SE of not less than 50 dB while maintaining a wideband response. SE shifts only 1% from the resonant frequency for both curved variations of the arc having radius $R_i$.

![Fig. 16. Simulated vs. measured normal and oblique SE of inward curved PFSS shield. (a) (c) TE-wave mode, (b) (d) TM-wave mode](image)

3) **Double curved PFSS shield**

Measured shielding performance of double curved shield is presented in Figs. 17(a) and 17(b). It may be observed that, SE shifts only 3% for both variations of $R_u$ and $R_s$. Moreover, a slight decrease in SE is also observed due to highly conformal surface. This configuration achieves SE of not less than 48 dB besides having highly conformal nature. The oblique incidence performance of this shield is illustrated in Figs. 17(c) and 17(d). The, SE varies only 3% from the resonant frequency when the angle of incidence increases up to 60°. Overall, this conformal configuration also exhibits stable incident angular response within acceptable limits.

![Fig. 17. Simulated vs. measured normal and oblique SE of double curved PFSS shield. (a) (c) TE-wave mode, (b) (d) TM-wave mode](image)

D. **Non-conformal vs. conformal**

The measured shielding performance of the proposed PFSS shield in flat and conformal configurations is compared in Fig. 18. It may be noticed that, SE is shifted less than 3% when PFSS is bent. These variations are attributed to curved profile and non-regular shape. However, this shift can be conveniently compensated at the design phase by considering the intended application. In general, this conformal shield exhibits stable shielding performance of not less than 55 dB and 48 dB for non-conformal/flat and conformal configurations respectively.

![Fig. 18. Simulated vs. measured normal and oblique SE of double curved PFSS shield. (a) (c) TE-wave mode, (b) (d) TM-wave mode](image)
E. Comparison with existing literature

Table IV gives a comparison of metrics for the proposed FSS with those available in the literature. From the comparison table, it is evident that, the most closely related works shows the good shielding performance and angular stability up to 45° either for flat, outward or inward curved configuration while maintaining low-profile design aspects. However, the higher degree of incidence angle tolerance is essential for the FSS and it is difficult to achieve over a highly conformal surface while achieving a good shielding performance and miniaturization alongside. The proposed FSS exhibits a higher order of angular stability up to 60° for highly conformal configurations including outward, inward and double curved. Moreover, relatively miniaturized single-layer unit cell structure, with good shielding characteristics are also the main advantages of the proposed FSS in comparison to the previously published works.

VII. CONCLUSION

This paper presented efficient and highly flexible PFSS based EM shields. The effectiveness of these shields is demonstrated for both planar and conformal configurations in order to serve a wide variety of applications. The design is simple in making and offers good shielding with at least 55 dB and 48 dB suppression at the resonant frequency for flat and all conformal configurations respectively. More importantly, a stable incident angular performance for both polarizations is also demonstrated for all variants under a wide range of incident angles up to 60°. The prototypes of all the proposed PFSS shields are fabricated and tested to verify their performance. SE of each configuration only varies a little as compared to simulations which can be priorly compensated at the design stage considering the intended application. The measurements are in line with the simulated full-wave analysis, hence making it a potential candidate for X-band applications, frequency selective radomes, high impedance surfaces (HISs), reduction of radar cross section of aerial vehicles and many others.

REFERENCES


Muhammad Bilal received his B.S degree in Telecommunication Engineering in 2011 from Baluchistan University of Information Technology, Engineering and Management Sciences (BUTECS), Quetta, Pakistan. He availed fully funded scholarship for Masters in Telecommunication Engineering in 2014 from University of Engineering and Technology (UET) Taxila, Pakistan. In 2018, he received the Ph.D. degree from the UET Taxila with specialization in Antennas and Electromagnetics. Since Feb 2018, he is perusing his career as an Assistant Professor with the Department of Telecommunication Engineering (BUTECS), where he is heading Higher Education Commission (HEC) funded research project while supervising graduate and post-graduate students. His current research interests include UWB-MIMO systems, High gain portable devices, Frequency Selective Surfaces and Reflectarrays. He has published in many reputable scientific journals and conferences

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