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# Resonant Tunnelling Diode Based High Speed Optoelectronic Transmitters

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## ABSTRACT

Resonant tunneling diode (RTD) integration with photo detector (PD) from epi-layer design shows great potential for combining terahertz (THz) RTD electronic source with high speed optical modulation. With an optimized layer structure, the RTD-PD presented in the paper shows high stationary responsivity of 5 A/W at 1310 nm wavelength. High power microwave/mm-wave RTD-PD optoelectronic oscillators are proposed. The circuitry employs two RTD-PD devices in parallel. The oscillation frequencies range from 20-44 GHz with maximum attainable power about 1 mW at 34/37/44GHz.

**Keywords:** resonant tunneling diode, photo diode, oscillator, photodetector

## 1. INTRODUCTION

As the demand for high speed wireless communication continues to increase exponentially, the need to utilize the high speed optic fiber which has been widely employed in most base stations for wireless broadband communications is higher than before. The resonant tunnelling diode photodetector (RTD-PD) technique can provide a low cost solution for the seamless integration of microwave/mm wave radio and fiber networks.

Resonant tunneling diode is considered to be the fastest solid-state electronic device. To date, the highest reported frequency of RTD oscillator is 1.92 THz [1]. To benefit from the RTDs' capability of high frequency sources, the monolithically integration of a light sensitive layer-photo detector was introduced with RTD epi-layer design. RTD based photo-detector shows great potential applications for optical-electrical modulation and photo-detection [3]. Optimization of RTD-PD layer structures to maximize the sensitivity and responsivity is one of the key factors for real applications.

Another key factor is high power/high frequency RTD-PD oscillator design. 30 Gbps wireless communication over a few centimeters distance has been demonstrated by using electrical-only modulated RTD oscillator at 500 GHz [2]. The short distance is mostly due to low output power which was in tens of microwatt range. Different from most RTD relaxation oscillator design [4][5], the proposed RTD-PD LC oscillator circuit utilize two RTD-PDs in parallel. Each device was biased individually to suppress the parasitic oscillation. Device size was also optimized in terms to provide best impedance matching. 1mW output power RTD-PD oscillators are reported here. It is expected that with improved device epi-layer design and oscillator output power level, RTD-PD technology can be widely utilized in the long distance/ ultra-high speed optical-wireless communication system.

## 2. RTD-PD LAYER STRUCTURE

As RTD devices exhibit negative differential conductance (NDC) gain over extra wide bandwidth, the RTD-PD device would act as optical receiver with a build in amplifier, which can extract from the optical modulated signal to the RF carriers and amplify it by using RTD as high frequency sources [3][6]. The typical RTD device layer structure consists of a narrow band gap - quantum well material sandwiched between two wide band gap-double barrier materials, namely double barrier quantum well (DBQW) structure. A RTD-PD structure integrates the DBQW within the photo-conductive layers as shown in Table 1, that InGaAs/AlAs RTD was monolithically integrated with 500 nm photo absorption layer-InGaAlAs on top of DBQW. The RTD-PD device is optimized to maximize the optical response, namely the responsivity and sensitivity. Responsivity is a critical parameter to determine the max output saturation power while sensitivity determines the lowest detectable power. The RTD-PD stationary state responsivity  $\mathfrak{R}(\lambda, V)$  at bias voltage  $V$  is defined

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by

$$\mathfrak{R}_{DC}(\lambda, V) = \frac{I_{light}(V) - I_{dark}(V)}{P_{light}(\lambda)} = M(V) \times \eta_{ph} \left( \frac{e\lambda}{hc} \right) \quad (1)$$

Where  $I_{light}(V)$  is the current illumination and  $I_{dark}(V)$  is the current under dark conditions,  $P_{light}(\lambda)$  is the incident optical power,  $\lambda$  is the wavelength of the incident light wave,  $M(V)$  accounts for the internal current gain effect (due to electron-holes generation and/or impact ionization mechanisms),  $\eta_{ph}$  is the external quantum efficiency;  $e$ ,  $h$  and  $c$  are the elementary charge, the Planck constant and the speed of light in the vacuum, respectively. The built-in gain factor  $M(V)$  accounts for the internal current gain effect (due to the photo-generated electron-holes pairs and impact ionization mechanisms). Ideally every photo produces one pair of electron-hole, which contribute to photo current without any loss, the quantum efficiency is equal to one and the ideal responsivity would be given simply by  $\mathfrak{R}_i = \frac{e\lambda}{hc}$ . The ideal maximum responsivity for wavelength 1310 nm and 1550 nm is 1.05 A/W and 1.25 A/W. In reality, many factors reduce the ideal responsivity value such as limited optical absorption efficient, finite absorption layer thickness (penetration depth), and optical reflection at the semiconductor surface, losses due to carrier recombination, and also the detector bandwidth limitation. The frequency response of the RTD-PD is not presented in the paper.

Table 1. RTD-PD epi-layer I.

Layer #	Thickness (Å)	Composition	Doping (cm <sup>-3</sup> )	Description
12	450	InGaAs	3E19 : Si	Collector
11	800	InGaAs	3E18 : Si	Collector
10	5000	InGaAs	5E16 : Si	Absorption
9	20	InGaAs	Un-doped	Spacer
8	17	AlAs	Un-doped	Barrier
7	47	InGaAs	Un-doped	Well
6	17	AlAs	Un-doped	Barrier
5	20	InGaAs	InGaAs	Spacer
4	200	InGaAs	2E16 : Si	Spacer
3	80	InAlAs	3E18 : Si	Emitter
2	4000	InGaAs	3E19 : Si	Emitter
1	200	InP	Un-doped	Buffer
Substrate SI : InP				

## 2.1 DC characteristics

In this paper, 10×10 μm<sup>2</sup> sized RTD-PD are investigated. The central optical window is rectangular shape. As shown in Fig 1 (a), the optical window is 3×8 μm<sup>2</sup> for the device design while Fig 1 (b) shows the fabricated device with clear-opened optical window. The I-V characteristics under dark and light illumination are compared in Fig 2. Under dark condition, the device shows negative differential conductance (NDC) region between 1.1V-1.4V. The peak current density was about 0.3mA/μm<sup>2</sup>, peak to valley current ratio (PVCR) was about 1.5. The device stationary optical response was investigated with non-modulated continuous wave (CW) light at 1310 nm and 1550 nm. The measurement work was done at Universidade do Algarve. The measurement setup is shown in the inset of Fig 2. The RTD-PD was biased from left (probe). The cleaved single mode optical fiber was aligned with the optical window from device top mesa. The light with wavelength 1310 nm and 1550 nm was coupled into the RTD-PD with 0.1mW and 1 mW power. As shown in Fig 2. (a), the optical DC response of 1310 nm laser with different optical power are compared with dark current. The maximum response was about 5 A/W when RTD-PD was biased close to the peak voltage and the injected light power was 1 mW. As shown in Fig 2. (b), the maximum responsivity of 1550 nm (1 mW) was about 2 A/W. The main difference is due to high absorption efficiency at 1310nm wavelength.

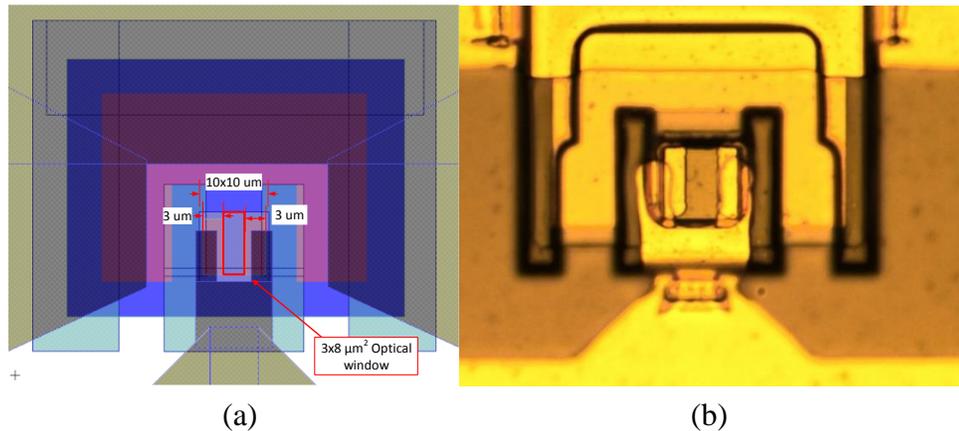


Figure 1. (a) RTD-PD layout design. The device size was  $10 \times 10 \mu\text{m}^2$ , the optical window size was  $3 \times 8 \mu\text{m}^2$ . (b) The micrograph of the top-view of the fabricated device. The optical window is clearly noted.

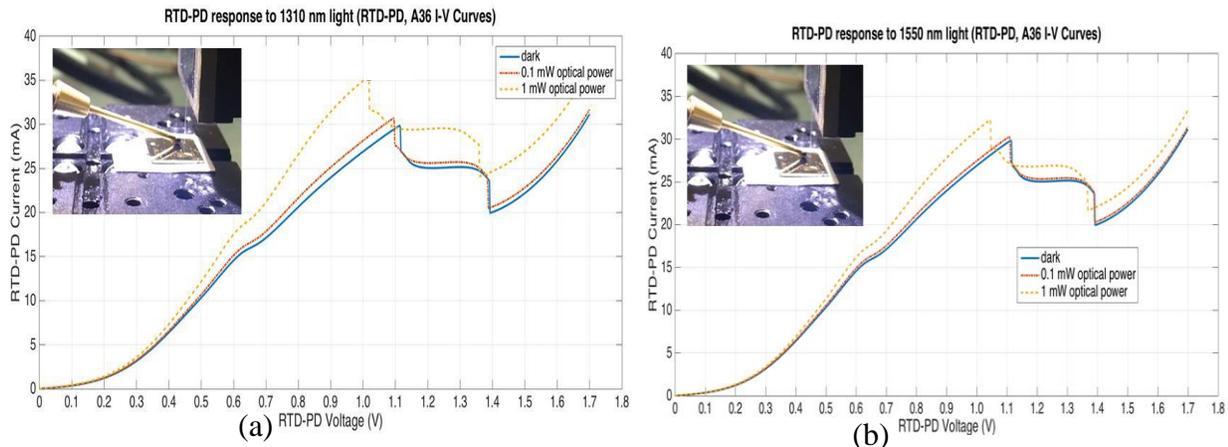


Figure 2. RTD-PD stationary optical response with light injection power of 0.1 mW and 1 mW (a) The optical response of 1310 nm wavelength. (b) The optical response of 1550 nm wavelength. The device shows large responsivity of 5 A/W for 1310 nm compared to 1550 nm due to high absorption efficiency. Inset shows the measurement setup. The device was biased from left (probe). The injected light was coupled from the top side.

### 3. RTD-PD OSCILLATORS

High frequency RTD-PD oscillator design utilized the device described above. New epi-layer structure including double light absorption layer on both side of DBQW as shown in Table 2 are introduced and characterized for the consideration of power performance and better optical control. As shown in Figure 3, two different sized devices,  $10 \times 10 \mu\text{m}^2$ ,  $15 \times 15 \mu\text{m}^2$  are investigated. The central optical windows are  $3 \times 8 \mu\text{m}^2$  and  $5 \times 10 \mu\text{m}^2$  respectively. The NDC region was between 1.2V-2.4V. The peak current density was about  $0.16 \text{mA}/\mu\text{m}^2$  and peak to valley current ratio (PVCR) was about 2. The measurement results are summarize in Table 3, where  $\Delta V$  is the peak-valley voltage difference,  $\Delta I$  is the peak-valley current difference. The maximum negative difference conductance  $G_n$  is given by  $\frac{3\Delta I}{2\Delta V}$ . The estimated maximum power for each single device is given by  $\frac{3}{16} \Delta V \Delta I$  when load conductance is  $G_L = G_n/2$  [7][8]. Benefit from large  $\Delta V$ , these devices shows potential for higher power oscillator design than the above epi-layer I (Table 1) devices.

Table 2. RTD-PD epi-layer II.

Layer #	Thickness (Å)	Composition	Doping (cm <sup>-3</sup> )	Description
13	1000	InGaAs	3E19 : Si	Collector
12	3000	InAlAs	3E18 : Si	Collector
11	5000	InGaAlAs	2E16 : Si	Absorption
10	20	InGaAlAs	Un-doped	Spacer
9	17	AlAs	Un-doped	Barrier
8	57	InGaAs	Un-doped	Well
7	17	AlAs	Un-doped	Barrier
6	20	InGaAlAs	Un-doped	Spacer
5	5000	InGaAlAs	2E16 : Si	Absorption
4	3000	InAlAs	3E18 : Si	Emitter
3	5000	InGaAs	3E19 : Si	Emitter
2	1000	InAlAs	Un-doped	
1	200	InP	Un-doped	Buffer

Substrate SI : InP

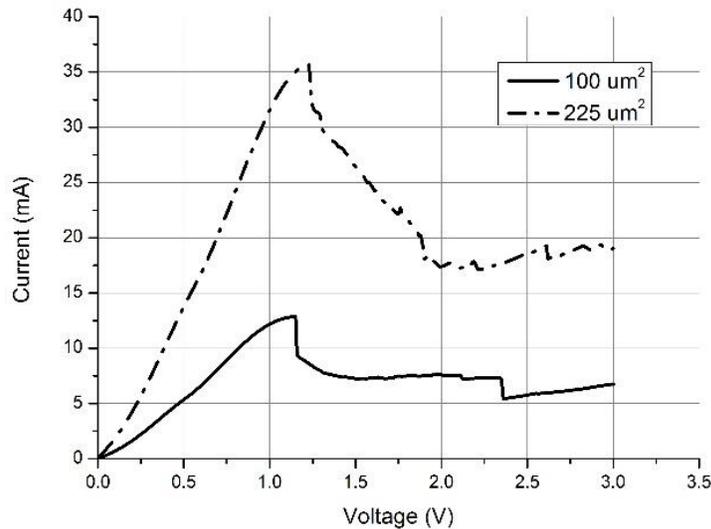


Figure 3. The measured RTD-PD IV characteristics (dark). Two different device size  $10 \times 10 \mu\text{m}^2$  and  $15 \times 15 \mu\text{m}^2$  are compared. The negative differential resistance region is between 1.2V~2.4V.

Table 3. RTD-PD DC characteristics.

Device size ( $\mu\text{m}^2$ )	Optical window ( $\mu\text{m}^2$ )	$\Delta V$ (V)	$\Delta I$ (mA)	$G_n$ (mS)	Estimated Power (mW)
100	3x8	1.2	7.4	9.3	1.7
225	5x10	1.2	18.5	23.1	3.5

Different from normal RTD-PD relaxation oscillator design [4][5][6] with a single RTD-PD device biased in series with a large inductor (transmission line), the proposed RTD-PD LC oscillator circuit utilize two RTD-PDs in parallel, as shown in Fig. 4(c). Each device is biased individually with shunt resistor to suppress the bias oscillation [9]. The schematic circuit of RTD-PD oscillator is as shown in Fig 4 (a), where  $R_e$  is the stabilizing resistor,  $C_e$  is the bypass capacitor,  $C_b$  is the DC block capacitor,  $R_L$  is the load. The RF equivalent circuit is shown in Fig 4(b). Each RTD-PD is presented by its negative differential conductance -  $G_{1n}/- G_{2n}$  in parallel with self-capacitances  $C_{1n}/C_{2n}$ . The oscillation frequency is determined by

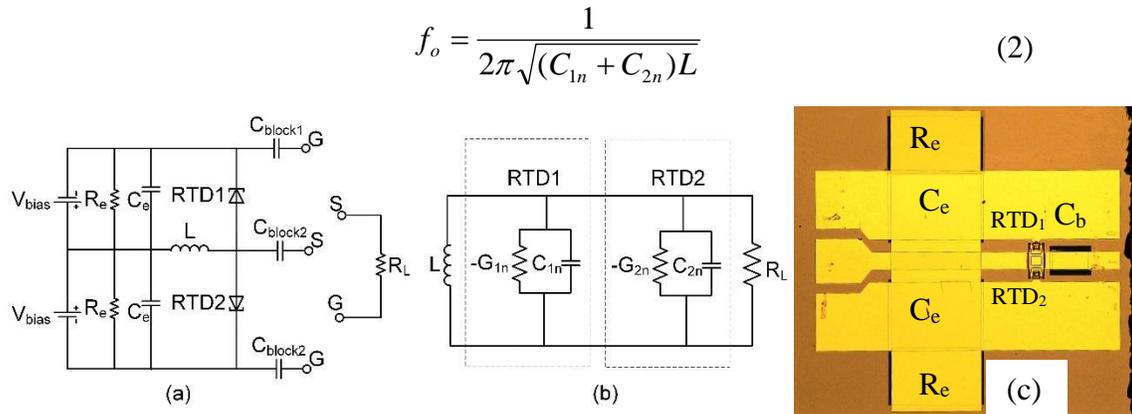


Figure 4. (a) The two RTD oscillator schematic circuit. Each RTD was biased individually with its own DC stabilization circuit  $R_e$  and  $C_e$ . (b) RTD oscillator RF equivalent circuit excluding parasitic elements. (c) Fabricated RTD-PD oscillator,  $R_e=20\ \Omega$ ,  $C_e=60\ \text{pF}$ ,  $C_b=7\ \text{pF}$ .

The fabricated oscillator photograph is shown in Fig 4 (c).  $R_e$  is thin film NiCr resistor with  $R_e=20\ \Omega$ ,  $C_e/C_b$  is metal-insulator-metal (MIM) capacitor with  $C_e=60\ \text{pF}$ ,  $C_b=7\ \text{pF}$ . The inductance is introduced by CPW structure. The inductor value is determined by the CPW length. The design procedure can be found in Ref. [9]. To effectively suppress the parasitic oscillation,  $R_e$  should satisfy that  $R_e < 1/G_n$ . From the impedance matching point of view, to obtain the maximum power, the load conductance should satisfy  $G_L = G_n/2$ . For a  $50\ \Omega$  load, the optimized device conductance is about  $40\ \text{mS}$ . Referring to Table 2, for  $225\ \mu\text{m}^2$  single device,  $G_n=23.1\ \text{mS}$ . By utilizing two  $225\ \mu\text{m}^2$  RTD-PD device in parallel, the total conductance is two times larger which is close to  $40\ \text{mS}$ . In this case,  $222\ \mu\text{m}^2$  RTD-PD oscillator provide better impedance matching, the output power is expected high. The measurement results shown in the following section validate the design.

#### 4. MEASUREMENT RESULTS

The oscillators were characterized on-wafer by using Agilent E4448A spectrum analyzer. A Cascade ACP65 RF probe was used to probe the output of these oscillators. As the cable loss/system loss is not accurately specified by the manufacturer, the actual power loss was characterized by calibrated another signal source Wiltron 68187B. The measurement results are summarized in Table 4. It shows that high frequency RTD-PD oscillators with over  $1\ \text{mW}$  output power have been developed. The frequency can be further increased by optimize the CPW design. The difference between designed and measured frequency is mostly due to the approximation of the device capacitance.

Table 4. RTD-PD oscillator measurement results.

Device size ( $\mu\text{m}^2$ )	Optical window ( $\mu\text{m}^2$ )	CPW length ( $\mu\text{m}$ )	Designed Freq(GHz)	Measured Freq(GHz)	Measured Power (dBm)	Calibrated Power (dBm)
100	24	700	30	20.7	-6.7	-1.7
100	24	260	60	34.4	-6.1	0.4
100	24	260	60	37.0	-7.1	0.1
225	50	75	90	44.4	-7.1	+1.4

#### 5. DISCUSSION AND CONCLUSION

The results presented in the paper show great potential to utilize RTD-PD as high speed optical modulated RF transmitter. The RTD-PD approach to optical/wireless data transmission provides a high efficiency, low cost solution for seamless integration of high speed fiber network with THz wireless communication. The increased power level makes it feasible for long distance applications. More work such as optical injection locking, optical-wireless data transmission, is currently being done with these RTD-PD oscillators.

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